

Design for Lyapunov Stability and Stabilization of Control Systems

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Modern Control System Representation

State equation

$$\frac{dx(t)}{dt} = Ax(t) + bu(t)$$

Output equation

$$y(t) = cx(t)$$

Linear system

$$\frac{dx(t)}{dt} = f(x(t), u(t))$$

$$y(t) = g(x(t), u(t))$$

Nonlinear system

A : system matrix

b : input matrix

c : output matrix

$x(t)$: state vector

$u(t)$: input vector

$y(t)$: output vector

Outline

- Review of Previous Lecture
- Design of Lyapunov Functions for Stability
Linear Matrix Inequality (LMI)
- Design of Lyapunov Functions for Stabilization
Bilinear Matrix Inequality (BMI)

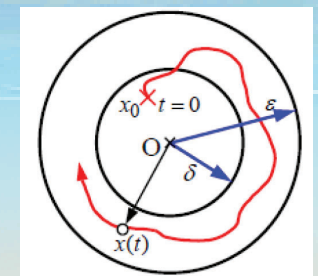
Lyapunov Stability Definitions (1)

Time-invariant autonomous (no control) system

$$\dot{x} = f(x), \quad x(0) = x_0, \quad f : \text{Lipschitz C.}$$

Equilibrium point $x_e \Leftarrow f(x_e) = 0$

Suppose $f(0) = 0 \Rightarrow x_e = 0$

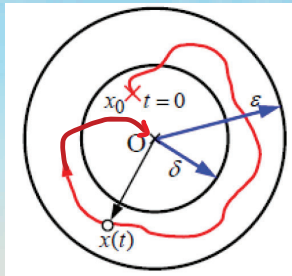


(1) The equilibrium $x_e = 0$ is **stable in the sense of Lyapunov**, if

$$\forall \epsilon > 0, \exists \delta > 0, \text{ s.t. } \|x_0\| \leq \delta \Rightarrow \|x(t)\| \leq \epsilon, \forall t \geq 0$$

Lyapunov Stability Definitions (2)

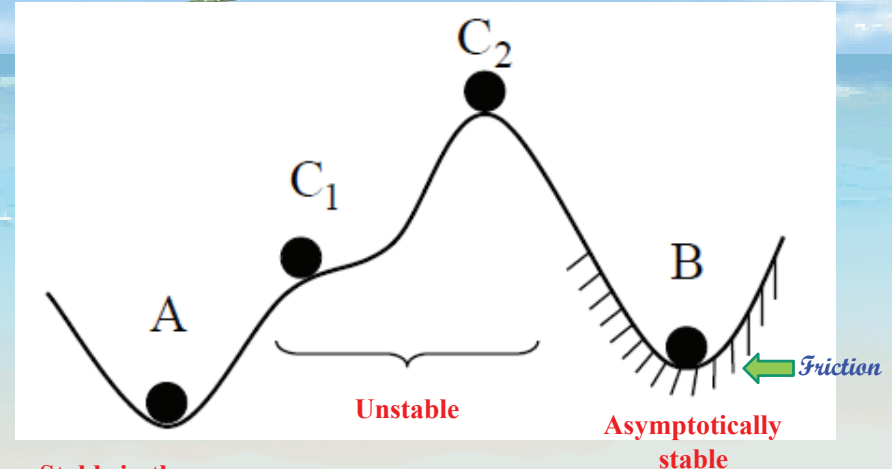
- (2) The equilibrium $x_e = 0$ is **asymptotically stable**, if it is stable, and $\exists \delta > 0$, s.t. $\|x_0\| \leq \delta \Rightarrow \lim_{t \rightarrow \infty} x(t) = 0$



Lyapunov Stable (LS)
 ↑
 Asymptotically Stable (AS)
 ↑
 Exponentially Stable (ES)

- (3) The equilibrium $x_e = 0$ is **exponentially stable**, if $\exists \delta > 0, c > 0, \lambda > 0$ s.t. $\|x(t)\| \leq c\|x_0\|e^{-\lambda t}, \forall \|x_0\| \leq \delta$

Lyapunov Stability Definitions (3)



Stable in the sense of Lyapunov

Asymptotically stable

Lyapunov Stability Theorem

If there is $V(x)$ such that

$$V(0) = 0 \text{ and } V(x) > 0, \forall x \neq 0$$

$$\dot{V}(x) \leq 0, \quad \forall x$$

then $x_e = 0$ is stable (in the sense of Lyapunov)

Moreover, if

$$\dot{V}(x) < 0, \quad \forall x \neq 0$$

then $x_e = 0$ is asymptotically stable

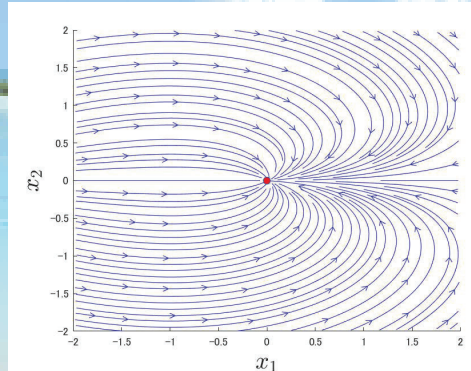
$V(x)$ is called a **Lyapunov function (candidate)**

Stability of Nonlinear Systems (1)

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -x_1 + 2x_2^2 \\ -x_1x_2 - x_2 \end{bmatrix}$$

$$V_3(x(t)) = x_1^2(t) + 2x_2^2(t)$$

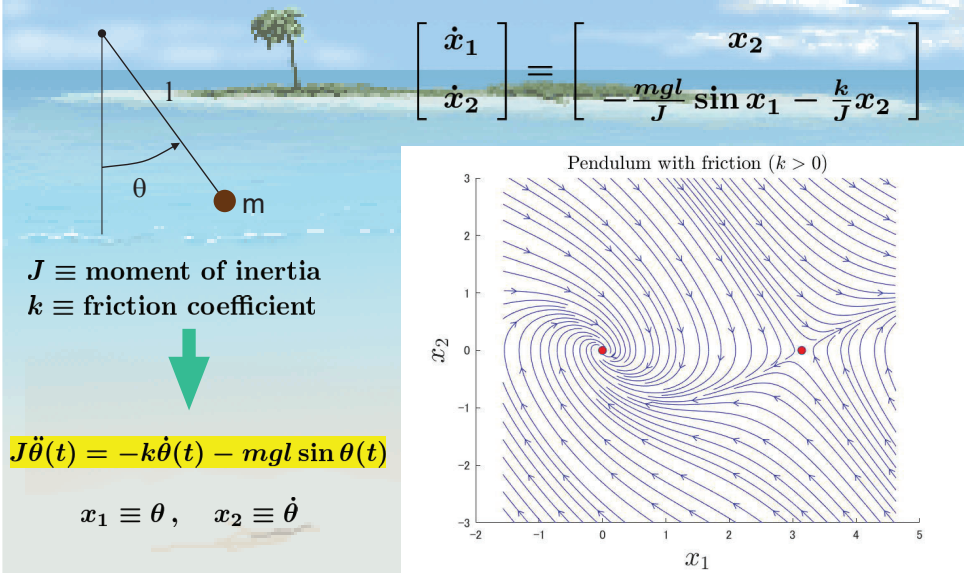
$$\begin{aligned} \frac{d}{dt} V_3 &= \frac{d}{dt} \{x_1^2 + 2x_2^2\} = 2x_1\dot{x}_1 + 4x_2\dot{x}_2 \\ &= 2x_1(-x_1 + 2x_2^2) + 4x_2(-x_1x_2 - x_2) \\ &= -2x_1^2 - 4x_2^2 = -2V_3 < 0, \quad \forall x \neq 0 \end{aligned}$$



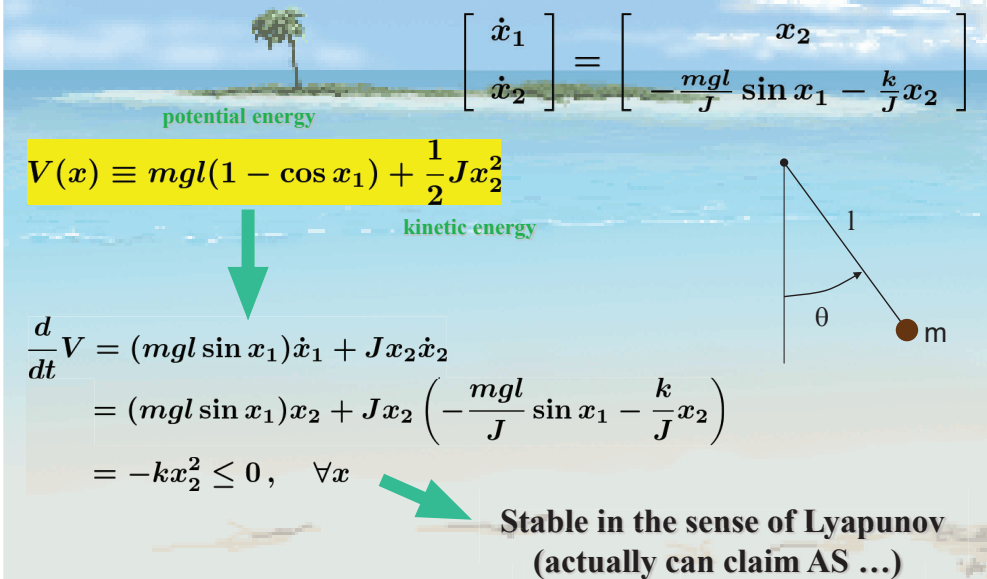
Asymptotically Stable (AS)

$$\lim_{t \rightarrow \infty} V_3(x(t)) = 0 \iff \lim_{t \rightarrow \infty} x(t) = 0, \quad \forall x(0)$$

Stability of Nonlinear Systems (2)



Stability of Nonlinear Systems (3)



Design of Lyapunov Functions for Stability (1)

NO almighty method for general **nonlinear** systems

YES when the systems are **linear** $\dot{x}(t) = Ax(t)$

Consider $V(x) = x^T P x$, $P > 0$

(P symmetric positive definite $\iff V(x)$ positive definite)

$$\begin{aligned} \implies \dot{V}(x) &= \dot{x}^T P x + x^T P \dot{x} \\ &= (Ax)^T P x + x^T P (Ax) \\ &= x^T (A^T P + P A) x \end{aligned}$$

($\dot{V}(x)$ negative definite $\iff A^T P + P A$ symmetric negative definite)

Asymptotically Stable (**AS**) $\iff A^T P + P A < 0$, $P > 0$

Design of Lyapunov Functions for Stability (2)

The linear system $\dot{x}(t) = Ax(t)$ is AS (ES) **iff**

(1) All real parts of A 's eigenvalues are negative

(2) $\forall Q > 0, \exists P > 0$ s.t. $A^T P + P A = -Q$

(3) $\exists P > 0$ s.t. $A^T P + P A < 0$

Lyapunov Equation

(4) $\exists X > 0$ s.t. $A X + X A^T < 0$

Linear Matrix Inequality (LMI)

Lyapunov equations and LMIs can be solved efficiently with MATLAB

Definition and Properties of LMI (1)

Notations

$$P > 0 \iff P = P^T : \text{positive definite}$$

$$P < 0 \iff P = P^T : \text{negative definite}$$

LMI formulation (in scalar variables)

$$F(x) = F_0 + x_1 F_1 + \dots + x_n F_n < 0$$

$$x = [x_1, \dots, x_n]^T, \quad F_i = F_i^T$$

LMI is convex

$$F(x) < 0, \quad F(y) < 0 \implies$$

$$F(\lambda x + (1 - \lambda)y) < 0, \quad \forall \lambda \in [0, 1]$$

LMI can be simultaneous

$$F_1(x) < 0, \quad F_2(x) < 0 \iff \begin{bmatrix} F_1(x) & 0 \\ 0 & F_2(x) \end{bmatrix} < 0$$

Definition and Properties of LMI (2)

LMI in matrix variable

$$P > 0, \quad A^T P + P A < 0$$

$$P = \begin{bmatrix} x_1 & x_2 \\ x_2 & x_3 \end{bmatrix} \quad P = x_1 \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} + x_2 \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} + x_3 \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix}$$

$E_1 \quad E_2 \quad E_3$



$$P > 0 \iff F_1(x) = x_1(-E_1) + x_2(-E_2) + x_3(-E_3) < 0$$

$$A^T P + P A < 0 \iff$$

$$F_2(x) = x_1(A^T E_1 + E_1 A) + x_2(A^T E_2 + E_2 A) + x_3(A^T E_3 + E_3 A) < 0$$

Definition and Properties of LMI (3)

Properties of positive (negative) definite matrices

$$P > 0 \iff \forall \lambda(P) > 0$$

$$P < 0 \iff \forall \lambda(P) < 0$$

$$P > 0 \iff W^T P W > 0 \quad (|W| \neq 0)$$

$$P < 0 \iff W^T P W < 0 \quad (|W| \neq 0)$$

Schur Complement

$$\begin{bmatrix} P_1 & P_2 \\ P_2^T & P_3 \end{bmatrix} > 0 \iff P_1 > 0 \quad \& \quad P_3 - P_2^T P_1^{-1} P_2 > 0$$

$$\iff P_3 > 0 \quad \& \quad P_1 - P_2 P_3^{-1} P_2^T > 0$$

Definition and Properties of LMI (4)

LMI: linear w.r.t. (matrix or scalar) variables

$$A^T P + P A < 0 \quad X A + A^T X < 0$$

$$\begin{bmatrix} A^T P + P A & Y B \\ B^T Y^T & -\gamma I \end{bmatrix} < 0$$

Not LMI but equivalent to LMI

$$P^2 > I \iff \begin{bmatrix} P & I \\ I & P \end{bmatrix} > 0 \quad P^2 < I \iff \begin{bmatrix} I & P \\ P & I \end{bmatrix} > 0$$

$$A^T P + P A + P D D^T P + E^T E < 0$$

$$\iff \begin{bmatrix} A^T P + P A + E^T E & P D \\ D^T P & -I \end{bmatrix} < 0$$

Solving LMIs in Robust Control Toolbox

$$P > 0, \quad A^T P + P A < 0$$

```
A=[0 1; -2 -3];
```

```
setlmis([])
```

```
p=lmivar(1,[2 1]);
```

```
lmiterm([1 1 1 p],1,A,'s');
```

```
lmiterm([1 2 2 p],[-1,1]);
```

```
lmis=getlmis;
```

```
[tmin,xfeas]=feasp(lmis)
```

```
if (tmin<0)
```

```
    P=dec2mat(lmis,xfeas,p)
```

```
end
```

$$A = \begin{bmatrix} 0 & 1 \\ -2 & -3 \end{bmatrix}$$

$$\begin{bmatrix} A^T P + P A & 0 \\ 0 & -P \end{bmatrix} < 0$$

$$P = \begin{bmatrix} 1.5143 & 0.2958 \\ 0.2958 & 0.3484 \end{bmatrix}$$

Extension to Stability with Convergence Rate

The linear system $\dot{x}(t) = Ax(t)$

$$\|x(t)\| < c\|x(0)\| e^{-\mu t} \quad (\mu > 0) \iff$$

(1) All real parts of A 's eigenvalues $< -\mu < 0$

(2) $\forall Q > 0, \exists P > 0$ s.t. $(A + \mu I)^T P + P(A + \mu I) = -Q$

(3) $\exists P > 0$ s.t. $A^T P + P A < -2\mu P$

Lyapunov Equation

(4) $\exists X > 0$ s.t. $AX + XA^T < -2\mu X$

Linear Matrix Inequality (LMI)

Extension to Robust Stability (1)

The uncertain linear system

$$\dot{x}(t) = Ax(t), \quad A = \lambda A_1 + (1 - \lambda)A_2, \quad \lambda \in [0, 1]$$

is AS for any λ if

$$\exists P > 0 \text{ s.t. } A_1^T P + P A_1 < 0, \quad A_2^T P + P A_2 < 0$$



$$\exists X > 0 \text{ s.t. } X A_1^T + A_1 X < 0, \quad X A_2^T + A_2 X < 0$$

Extension to Robust Stability (2)

The uncertain linear system

$$\dot{x}(t) = (A + DF(t)E)x(t), \quad \|F(t)\| \leq 1$$

is AS for any $F(t)$ if

$$\exists P > 0 \text{ s.t. } (A + DF(t)E)^T P + P(A + DF(t)E) < 0$$



$$A^T P + P A + P D D^T P + E^T E < 0$$



$$\begin{bmatrix} A^T P + P A + E^T E & P D \\ D^T P & -I \end{bmatrix} < 0$$

Design of Lyapunov Functions for Stabilization

Consider state feedback for linear control system

$$\dot{x}(t) = Ax(t) + Bu(t)$$

$$u(t) = Kx(t)$$

Closed-loop system

$$\dot{x}(t) = (A + BK)x(t)$$

Closed-loop system is AS **iff**

$$\exists P > 0 \text{ s.t. } (A + BK)^T P + P(A + BK) < 0$$

BMI with respect to P, K due to PBK can not be solved with MATLAB

Design of Lyapunov Functions for Stabilization

Closed-loop system

$$\dot{x}(t) = (A + BK)x(t)$$

Closed-loop system is AS **iff**

$$\exists X > 0 \text{ s.t. } (A + BK)X + X(A + BK)^T < 0$$

$$\updownarrow \quad KX = M$$

$$\exists X > 0, M \text{ s.t. } AX + BM + (AX + BM)^T < 0$$

LMI with respect to X, M

$$\downarrow \quad K = MX^{-1}$$

Extension to Robust Stabilization (1)

State feedback for the uncertain linear control system

$$\dot{x}(t) = Ax(t) + Bu(t), \quad u(t) = Kx(t)$$

$$A = \lambda A_1 + (1 - \lambda)A_2, \quad \lambda \in [0, 1]$$

Control specification

$$\|x(t)\| < c\|x(0)\| e^{-\mu t} \quad (\mu > 0) \quad \forall \lambda \in [0, 1]$$

Closed-loop system

$$\dot{x}(t) = (A + BK)x(t)$$

$$\begin{aligned} A + BK &= \lambda A_1 + (1 - \lambda)A_2 + BK \\ &= \lambda(A_1 + BK) + (1 - \lambda)(A_2 + BK) \end{aligned}$$

Extension to Robust Stabilization (2)

Closed-loop system

$$\dot{x}(t) = (A + BK)x(t)$$

$$A + BK = \lambda(A_1 + BK) + (1 - \lambda)(A_2 + BK)$$

is AS and $\|x(t)\| < c\|x(0)\| e^{-\mu t} \quad (\mu > 0) \quad \forall \lambda \in [0, 1]$ if

$$\updownarrow$$

$$\exists X > 0 \text{ s.t. } X(A_1 + BK)^T + (A_1 + BK)X < -2\mu X$$

$$X(A_2 + BK)^T + (A_2 + BK)X < -2\mu X$$

$$\updownarrow \quad KX = M$$

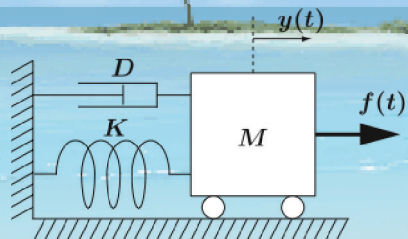
LMIs

$$A_1 X + BM + (A_1 X + BM)^T + 2\mu X < 0$$

$$A_2 X + BM + (A_2 X + BM)^T + 2\mu X < 0$$

Robust Stabilization for MDK Systems (1)

MDK (mass-spring-damper) system



Mass $M = 1$
Spring $K \in [2, 3]$
Damper $D = 3$

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -\frac{K}{M} & -\frac{D}{M} \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{M} \end{bmatrix} u(t)$$

$$A = \begin{bmatrix} 0 & 1 \\ -K & -3 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

Robust Stabilization for MDK Systems (2)

Control Specification

$$\|x(t)\| < c\|x(0)\| e^{-\mu t} \quad (\mu = 3) \quad \forall K \in [2, 3]$$

$$K = 2\lambda + 3(1 - \lambda), \quad \lambda \in [0, 1]$$

$$A = \lambda A_1 + (1 - \lambda) A_2$$

$$A_1 = \begin{bmatrix} 0 & 1 \\ -2 & -3 \end{bmatrix}, \quad A_2 = \begin{bmatrix} 0 & 1 \\ -3 & -3 \end{bmatrix}$$

$$X > 0 \quad \begin{aligned} A_1 X + B M + (A_1 X + B M)^T + 2\mu X &< 0 \\ A_2 X + B M + (A_2 X + B M)^T + 2\mu X &< 0 \end{aligned}$$

Robust Stabilization for MDK Systems (3)

$A1=[0 \ 1; -2 \ -3]; \quad A2=[0 \ 1; -3 \ -3];$
 $B=[0; 1]; \quad \mu=3;$

```
setlmis([])
x=lmivar(1,[2 1]);
m=lmivar(2,[1 2]);
```

```
lmiterm([1 1 1 x],A1+mu*eye(2),1,'s');
lmiterm([1 1 1 m],B,1,'s');
lmiterm([1 2 2 x],A2+mu*eye(2),1,'s');
lmiterm([1 2 2 m],B,1,'s');
lmiterm([1 3 3 x],-1,1);
```

**Robust Control Toolbox
in MATLAB**

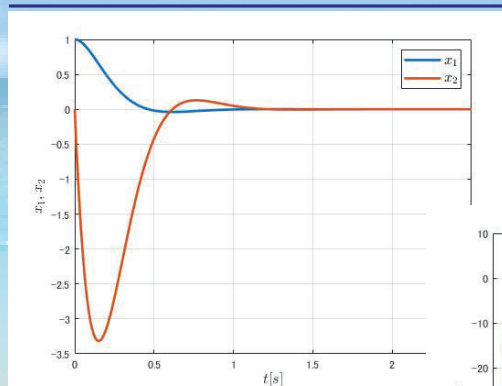
```
lmis1=getlmis;
[tmin,xfeas]=feasp(lmis1)
```

```
if (tmin<0)
    X=dec2mat(lmis1,xfeas,x)
    M=dec2mat(lmis1,xfeas,m)
    K=M*inv(X)
```

end

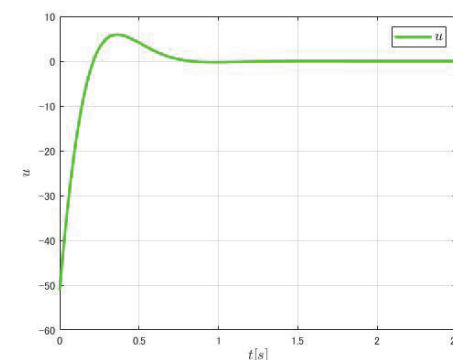
$$X = 10^3 * \begin{bmatrix} 0.0410 & -0.1873 \\ -0.1873 & 1.3530 \end{bmatrix} \quad \begin{aligned} M &= \begin{bmatrix} -675.4848 & -669.1301 \end{bmatrix} \\ K &= \begin{bmatrix} -51.0492 & -7.5620 \end{bmatrix} \end{aligned}$$

Robust Stabilization for MDK Systems (4)



State Convergence

Control Input



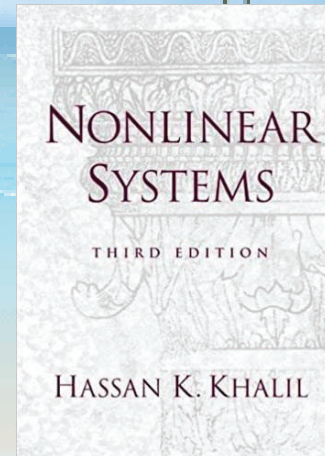
Conclusion

- Review of Lyapunov stability in modern control systems
- Definition of LMIs and properties
- LMIs for stability analysis of linear systems
- LMIs for stabilization of linear systems

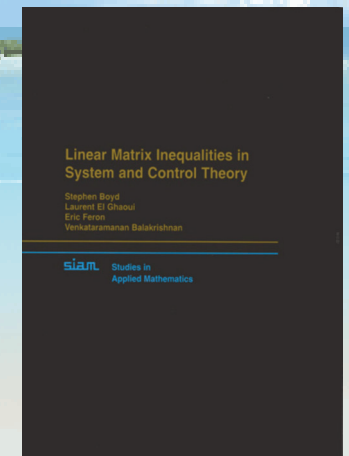
• More Topics:

Controller design with structure specification/limitation
Extension to LaSalle's Invariance Principle, etc

References



(1) H. K. Khalil: Nonlinear Systems.
Prentice-Hall, New Jersey, 1996.



(2) S. Boyd et al: Linear Matrix Inequalities in
Systems and Control Theory, SIAM, 1994

End

terima kasih banyak

Thank you for your kind attention!

Additional References

- Linearized nonlinear systems

$$\dot{x}(t) = f(x(t)) \implies \dot{x}(t) = \left(\frac{\partial f}{\partial x} \bigg|_{x=0} \right) x + O(x^2)$$

- Time-variant systems

$$\dot{x}(t) = A(t)x(t)$$

- Uncertain systems (norm-bounded or polytopic)

$$\dot{x}(t) = (A + \Delta A(t, x)) x(t)$$

$$\dot{x}(t) = \left(\sum_{i=1}^N \mu_i A_i \right) x(t), \quad \mu_i \geq 0, \sum_{i=1}^N \mu_i = 1$$

- Stochastic control systems

$$dx(t) = [Ax(t) + Bu(t)] dt + Hx(t) dw(t)$$